VARIABLE FREQUENCY SWITCHING AMPLIFIER AND METHOD THEREFOR

Field of the Invention

5   The present invention relates generally to switching amplifiers and more specifically to a variable frequency switching amplifier.

Related Art

Switching amplifiers, such as, for example, class D audio amplifiers have
10   associated with them a potentially high level of radiated energy at multiples of the switching frequency. This radiated switching energy can result in electro-magnetic interference (EMI) issues in adjacent circuitry. This can be particularly troublesome when trying to package the switching amplifier with a radio tuner, such as in an audio-video (AV) receiver. Currently, careful layout
15   and filter design, coupled with shielding, is used to reduce the radiation. For example, the tuner is placed at a far enough distance from the switching amplifier so as to reduce the effects of radiation. Also, heavy shielding may be used to minimize the radiation. However, these solutions become costly and some switching energy, regardless of placement, shielding, or filtering will still
20   be present. Therefore, a need exists for a low cost method for enabling a switching amplifier to be in proximity to a tuner without corrupting the received signal.

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Brief Description of the Drawings

The present invention is illustrated by way of example and not limitation in the accompanying figures, in which like references indicate similar elements, and in which:

FIG. 1 illustrates, in block diagram form, a switching amplifier system in accordance with one embodiment of the present invention;

FIG. 2 illustrates a portion of a frequency spectrum of the switching energy of an amplified switching signal in accordance with one embodiment of the present invention;

FIG. 3 illustrates an AM frequency band in accordance with one embodiment of the present invention; and

FIG. 4 illustrates, in block diagram form, a portion of an integrated circuit in accordance with one embodiment of the present invention.

Skilled artisans appreciate that elements in the figures are illustrated for simplicity and clarity and have not necessarily been drawn to scale. For example, the dimensions of some of the elements in the figures may be exaggerated relative to other elements to help improve the understanding of the embodiments of the present invention.

Detailed Description

As discussed above, switching audio amplifiers, such as, for example, class D audio amplifiers, emit potentially high levels of radiated energy at

5   multiples of the switching frequency that can result in EMI issues with adjacent circuitry. For example, this adjacent circuitry may include a radio tuner. Therefore, problems may arise when packaging a radio tuner (such as, for example, an audio-video (AV) receiver) with a switching amplifier. For example, switching amplifiers typically switch in the 350 to 400 kHz range.

10   Therefore, second, third, and fourth harmonics of this frequency fall directly in the AM radio band (e.g. 530 to 1710 kHz). Higher order harmonics can also fall in the FM Intermediate Frequency (IF) band (e.g. 10.7MHz). (Note that the AM radio band and the FM IF band may be within different frequency ranges than those given above.)   Prior solutions have therefore attempted to solve

15   this problem through the use of layout, filter design, and shielding. However, these solutions become costly and increase the resulting product size. Therefore, a need exists for a lower cost solution that allows for cohabitation of a switching amplifier with a tuner that reduces interference caused by the switching energy.

20   FIG. 1 illustrates, in block diagram form, a switching amplifier system 10 in accordance with one embodiment of the present invention. Switching amplifier system 10 includes tuner 12 coupled to antenna 14, signal processing unit 16, and switching frequency controller 18. Switching frequency controller 18 is coupled to signal processing unit 16. Signal processing unit 16 and

25   switching frequency controller 18 are both coupled to receive a CLOCK signal which is provided, in one embodiment, by a crystal oscillator (not shown).

Signal processing unit 16 and switching frequency controller 18 are included within integrated circuit (IC) 32. IC 32 may include other components of FIG. 1 or other components not shown in FIG. 1. Signal processing unit 16 is coupled to switching amplifier 20 which is coupled to an optional low pass

5   filter 22 which is coupled to speaker 24. In one embodiment, tuner 12 is an AM-FM tuner used in proximity with switching amplifier 20. (Switching amplifier 20 may be a switching power amplifier, such as, for example, a switching class D audio power amplifier, or any type of switching differential output amplifier. Although FIG. 1 illustrates switching amplifier 20 as an

10   inverter, switching amplifier 20 may be designed in a variety of different ways and include additional circuitry components.) Therefore, in one embodiment, tuner 12 may be a car stereo that is located in the car dashboard in close proximity with the switching amplifier, also located in the dashboard. In alternate embodiments, tuner 12 may be an AM-only tuner. Tuner 12 may

15   receive a digital signal via antenna 14, such as a digital broadcast, or, in another embodiment, tuner 12 may receive an analog radio signal via antenna 14. In yet another embodiment, tuner 12 may receive both digital and analog radio signals via antenna 14, either independently or simultaneously.

In operation, tuner 12 receives a modulated radio signal from antenna 14,

20   and provides a demodulated audio signal 30, corresponding to the received modulated radio signal, to signal processing unit 16. Tuner 12 also provides a control signal 34 to switching frequency controller 18. (Note that in some embodiments control signal 34 may transfer more than one control signal to switching frequency controller 18. In this case, control signal 34 may be a

25   control bus having a plurality of conductors.) Switching frequency controller 18 provides a switching frequency indicator 31 to signal processing unit 16

which corresponds to the switching frequency, F_s, of switching amplifier 20.
Switching frequency controller 18 is therefore capable of modifying F_s in
response to control signal 34, as will be described further below in reference to
FIGs. 2-4. For example, control signal 34 may indicate to switching frequency

5   controller 18 the frequency at which frequency tuner 12 is tuned. Signal
processing unit 16 receives switching frequency indicator 31 and demodulated
audio signal 30 and generates an appropriately modulated signal, switching
signal 28, to drive switching amplifier 20. For example, signal processing unit
16 may generate a pulse width modulated (PWM) signal from demodulated

10   audio signal 30 to drive switching amplifier 20. This PWM signal may either
be a single-sided PWM signal (meaning only the rising or the falling edge is
modulated) or a two-sided PWM signal (meaning that both the rising and the
falling edges are modulated). Alternatively, signal processing unit 16 may
generate a pulse density modulated (PDM) signal instead of a PWM signal.

15   Therefore, signal processing unit 16 may use various different methods to
appropriately modify demodulated audio signal 30 and generate switching
signal 28 for driving switching amplifier 20. However, for ease of explanation
in the description to follow, it will be assumed that signal processing unit 16
generates a single-sided PWM signal as switching signal 28.

20   Switching amplifier 20 is illustrated in FIG. 1 as an inverter which
receives switching signal 28 and outputs amplified switching signal 26.
(However, as discussed above, switching amplifier 20 may be designed in a
variety of different ways.) Switching amplifier 20 switches at a frequency of
F_s. The amount of amplification depends generally on the magnitude of voltage

25   supply 21. However, the amplification of switching signal 28 also amplifies
any switching energy associated with switching signal 28 such that the

amplified switching energy within amplified switching signal 26 may cause interference in nearby circuitries, such as, for example, tuner 12. This interference path may be radiated (e.g. electro-magnetic radiation) or conducted.

5   Amplified switching signal 26 may be filtered through a filter such as low pass filter 22 of FIG. 1. Low pass filter 22 helps reduce the switching energy within amplified switching signal 26. The signal from low pass filter 22 is then output to speaker 24. (However, in alternate embodiments, no filtering may be performed on amplified switching signal 26. In these embodiments,
10   low pass filter 22 would not be present and amplified switching signal 26 can be output to speaker 24.)

FIG. 2 illustrates a portion of a frequency spectrum of the switching energy of amplified switching signal 26. The switching energy includes components centered at the switching frequency, F_s, as well as at the harmonics
15   of F_s. The switching energy also includes sidebands centered about each harmonic of F_s. These sidebands are due to intermodulation between F_s and its harmonics and the frequency, F_m, of the input audio signal (e.g. demodulated audio signal 30). The switching energy components centered at F_s and the harmonics of F_s can therefore be defined as $(n*F_s \pm m*F_m)$ where $n = 1, 2, 3 \ldots$
20   and $m = 1, 2, 3 \ldots$ For example, illustrated in FIG. 2 is a switching energy associated with an F_s of 375 kHz. Therefore, a first harmonic of F_s occurs at 375 kHz. The switching energy component centered about 375kHz is therefore $375 \pm m*F_m$. As the harmonics of F_s increase, the width of the non-negligible sideband portions (attributed to $m*F_m$) also increase, as can be seen in FIG. 2.
25   The second switching energy component is centered about the second harmonic

of F_s, 750kHz, the third centered about 1125kHz, and the fourth centered about 1500kHz.

Referring to FIG. 2, the AM band is defined as extending from 530 kHz to 1710kHz. Therefore, the second, third, and fourth switching energy

5   components (attributable to the harmonics of F_s and the non-negligible sidebands at each harmonic) fall within the AM band. This may therefore result in interference when trying to tune to an AM frequency that falls within these switching energy components. For example, if tuning to a tuning frequency such as 1120kHz on an AM radio dial coupled to a switching amplifier having a

10   switching frequency of 375kHz, interference from the switching amplifier may corrupt signal reception. In order to avoid the interference resulting from a tuning frequency falling within a switching energy component, a second switching frequency may be introduced. (Therefore, for illustration purposes, the 375kHz will be referred to as the first switching frequency.)

15   As illustrated in FIG. 2, a second switching frequency of 352.9kHz results in switching energy components centered about its harmonics that are spaced apart from those switching energy components attributable to the first switching frequency of 375kHz. Therefore, a first harmonic of the second switching frequency occurs at 352.9 kHz. The switching energy component

20   centered about 352.9kHz is therefore $352.9 \pm m*F_m$, as was described above. As the harmonics increase, the width of the non-negligible sideband portions (attributed to $m*F_m$) also increase, as can be seen in FIG. 2. The second switching energy component is centered about the second harmonic of the second switching frequency, 705.8kHz, the third centered about 1058.7kHz,

25   and the fourth centered about 1411.6kHz. Note that the switching energy components of each of the two switching frequencies (375kHz and 352.9kHz)

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do not overlap (as can be seen in FIG. 2). Therefore, in order to avoid interference, if a tuning frequency approaches a switching energy component of the first switching frequency (375kHz), the switching frequency can be shifted to the second switching frequency (352.9kHz).

5   FIG. 3 illustrates an AM frequency band in accordance with one embodiment of the present invention. In this embodiment, the first switching frequency, 375kHz, is generally used within the AM frequency band (530kHz to 1710kHz). However, within those bands where interference may occur (generally centered about the harmonics of the first switching frequency), the

10   switching frequency of switching amplifier 20 is shifted to the second switching frequency, 352.9kHz. For example, the second harmonic of 375kHz occurs at 750kHz. Therefore, if the tuning frequency falls within the range of 730kHz and 770kHz, the switching frequency of switching amplifier 20 is shifted from 375kHz to 352.9kHz. In this manner, the problematic range of 375kHz $\pm$ m*F_m

15   is avoided (for the example illustrated in FIGs. 2 and 3). Likewise, if the tuning frequency received from tuner 12 falls within 1100kHz and 1150kHz or within 1480kHz and 1520kHz, the switching frequency of switching amplifier 20 is shifted from 375kHz to 352.9kHz.

       Note that the ranges in FIG. 3 of 730kHz to 770kHz and 1480kHz to

20   1520kHz can be defined by n*F_s $\pm$ 20kHz, where F_s is the first switching frequency of 375kHz and n corresponds to the harmonic of F_s (where n=2 for the range of 730kHz to 770kHz and n=4 for the range of 1480 to1520). However, note that the range of 1100kHz to 1150kHz is n*F_s $\pm$ 25kHz. The $\pm$ 25kHz is chosen due to the AM band frequencies being broadcast at intervals of

25   10kHz. Therefore, if only $\pm$ 20kHz is used, the range would occur between 1105kHz and 1145kHz which are not centered about valid AM broadcast

frequencies. In alternate embodiments, the AM band frequencies may be broadcast at other intervals such as 5kHz or 20kHz, and therefore, different ranges may be defined other than the ± 20kHz or ± 25kHz used in the example of FIG. 3.

5   While the ranges of 730kHZ to 770kHz, 1100kHz to 1150kHz, and 1480kHz to 1520kHz are illustrated in FIG. 3, one of ordinary skill in the art can appreciate that different ranges may be defined that prevent the tuned frequency from falling within a switching energy component. For example, the ranges may be larger than those shown in FIG. 3. Also, alternate embodiments

10   may define more than just two switching frequencies at which to operate switching amplifier 20. For example, a third switching frequency may be used in addition to the two discussed above. Therefore, other ranges may be defined in which the third switching frequency would be used rather the first or second switching frequencies. More ranges can therefore be defined to indicate when

15   the various switching frequencies are used. Furthermore, alternate embodiments may utilize a feedback method to detect interference and use an algorithm to determine the new switching frequency (as will be described in more detail below).

Additionally, alternate switching frequencies may be used other than the

20   375kHz and 352.9kHz described in the examples above. Generally, the switching frequencies are chosen such that the first harmonics of the switching frequencies fall outside the AM frequency band. Therefore, switching frequencies within the range of 350kHz to 400kHz are typically used since the first harmonic of these frequencies fall below 530kHz. Also, the switching

25   frequencies may be chosen such that the harmonics of the switching frequencies do not fall within the AM or FM intermediate frequencies (IF). For example,

the AM IF may be defined at 450kHz, which is outside the range of 350kHz to 400kHz. Also, the FM IF may be defined at 10.7MHz, meaning a switching frequency within 350kHz and 400kHz may be chosen such that its harmonics do not fall within a predetermined band centered about the FM IF (10.7MHz).

5   Although the typical range is defined as 350kHz to 400kHz, a variety of switching frequencies outside this range may be used as well, depending on the needs of the switching amplifier system.

FIG. 4 illustrates, in block diagram form, a portion of integrated circuit (IC) 32 of FIG. 1, according to one embodiment of the present invention. FIG. 10   4 includes a portion of signal processing unit 16 which includes an analog-to-digital (A/D) converter 41 coupled to a sample rate converter 40 coupled to a processor 42. FIG. 4 also includes a portion of switching frequency controller 18 which includes divider 44 coupled to sample rate converter 40, and logic 46 coupled to divider 44. A/D converter 41 receives demodulated audio signal 30 15   and input sample rate 48 and provides a sampled demodulated audio signal 43 where the sample rate of A/D converter 41 is input sample rate 48. Sample rate converter 40 receives sampled demodulated audio signal 43 and input sample rate 48, and processor 42 outputs switching signal 28. Logic 46 receives control signal 34, and divider 44 receives the CLOCK signal and provides 20   switching frequency indicator 31 to sample rate converter 40.

In operation, A/D converter 41 may not be capable of providing the desired output sample rate which is indicated by switching frequency indicator 31. For example, if the switching frequency, F_s, of switching amplifier 20 is 375kHz, the desired output sample rate at the output of sample rate converter 40 25   may be 750kHz (which is a function of F_s where F_s is indicated by switching frequency indicator 31). However, A/D converter 41 may not be capable of

providing an output having a sampling frequency of 750kHz. In this situation, sample rate converter 40 may include an upsampler which receives the sampled demodulated audio signal 43 from A/D converter 41 and modifies its sample rate to achieve the desired 750kHz and provide this modified signal to

5   processor 42. Note that 375kHz and 750kHz are only examples used for illustration purposes; therefore, alternate embodiments may use other switching frequencies. In an alternate embodiment, A/D converter 41 may be capable of outputting sampled demodulated audio signal 43 at the desired output sample rate. In this embodiment, sample rate converter 40 is optional and may be left

10   out of integrated circuit 32. Processor 42 therefore receives the output from sample rate converter 40 (or A/D converter 41 if sample rate converter 40 is not there) and generates switching signal 28, which, in one embodiment, is a PWM signal, as discussed above.

Logic block 46 receives control signal 34, which, in the example illustrated in FIG. 4, is the tuning frequency from tuner 12. Logic 46 determines

15   whether the tuning frequency of tuner 12 falls within a range that requires a switching frequency shift. For example, referring to FIG. 3, logic 46 may determine when the tuning frequency falls within 730kHz and 770kHz, or within 1100kHz and 1150kHz, or within 1480kHz and 1520kHz. If logic 46

20   detects that a switching frequency shift is required, it provides an indicator to divider 44 which in turn provides switching frequency indicator 31 which indicates to signal processing unit 16 when a switching frequency shift is required. Therefore, in the embodiment illustrated in FIG. 4, divider 44 receives the CLOCK signal from the reference crystal oscillator and divides the

25   CLOCK frequency by a predetermined number which corresponds to the desired switching frequency. For example, if the CLOCK frequency is

designed to be 48MHz and the switching frequency is desired to be 375kHZ, logic block 46 may indicate a value of N to be 128, such that the CLOCK frequency/N (48MHz/128) indicates a switching frequency of 375kHz. Similarly, if the switching frequency is to be shifted to 352.9kHz, logic block

5   46 may indicate the value of N to be 136, such that switching frequency indicator 31 indicates 48MHz/136, which corresponds to 352.9kHz. Alternate embodiments may use other processing blocks in place of logic block 46 and divider 44 to detect the need for a switching frequency shift and provide the proper indication (e.g. via switching frequency indicator 31) to signal

10   processing unit 16 that results in the desired switching frequency shift.

In an alternate embodiment, demodulated audio signal 30 may be received as a sampled digital signal from a digital broadcaster received by antenna 14. In this embodiment, A/D converter 41 would not be needed since demodulated audio signal 30 would be a sampled demodulated audio signal that

15   could be provided directly to sample rate converter 40. Sample rate converter 40 would therefore receive input sample rate 48 and modify the sample rate of demodulated audio signal 30 to provide the desired switching frequency as indicated by switching frequency indicator 31.

Referring back to FIG. 1, control signal 34 may include a plurality of

20   control signals from tuner 12 to switching frequency controller 18. For example, one of these control signals may indicate that tuner 12 is in a seek or scan mode. In this case, switching frequency controller 18 may indicate, via switching frequency indicator 31, to signal processing unit 16 to stop switching signal 28 from switching during the seek or scan mode. This temporarily

25   prevents switching amplifier 20 from generating the amplified switching energy

components discussed above with reference to FIG. 2, thereby preventing tuner 12 from falsely locking on the switching energy components.

In yet another alternate embodiment, FIG. 1 may include a feedback system for detecting interference. As discussed above, switching amplifier system 10 may be a feedforward system that detects when the tuning frequency is in a predetermined range, and in response to this detection, cause the switching frequency to be shifted or modified. However, FIG. 1 may be designed to function in a feedback system where signal processing unit 16 may detect tuner interference based on a feedback signal, such as, for example, amplified switching signal 26, and in response to the detection of interference, provide a control signal (not shown) to switching frequency controller 18 to modify the switching frequency. Therefore, the switching frequencies may be modified in response to the detection of interference rather than in response to the tuner frequency falling within predetermined ranges. Also note that switching frequency controller 18 may modify the switching frequency by selecting between predetermined switching frequencies or by algorithmically modifying the switching frequency based on a previous switching frequency.

Note that the embodiments discussed above in reference to FIG. 1 may apply to both the digital and analog domains. For example, generation of the PWM signal (switching signal 28) can be either in the digital or analog domain. Also note that while FIG. 1 illustrates a single tuner (tuner 12) coupled to a switching amplifier (switching amplifier 20), switching energy components may be introduced by any adjacent switching amplifier. The modification of the switching frequency in response to tuner 12, as discussed above, would therefore apply regardless of the audio source of the switching amplifier. For example, the tuner may drive a low power headphone amplifier within a same

system where another driving source such as a CD or DVD player drives a switching amplifier. The switching amplifier driven by the other driving source may be close enough to the tuner to still cause interference. Therefore, the modification of the switching frequency of the switching amplifier, as described

5   above, would help reduce interference of the tuner caused by the switching energy components of the nearby switching amplifier.

In the foregoing specification, the invention has been described with reference to specific embodiments. However, one of ordinary skill in the art appreciates that various modifications and changes can be made without

10   departing from the scope of the present invention as set forth in the claims below. For example, the boxes illustrated in FIGs. 1 and 4 may be separated into more units or be combined into existing units. Furthermore, the blocks illustrated in FIGs. 1 and 4 may be implemented in hardware, software, or in a combination of hardware and software. Accordingly, the specification and

15   figures are to be regarded in an illustrative rather than a restrictive sense, and all such modifications are intended to be included within the scope of the present invention.

Benefits, other advantages, and solutions to problems have been described above with regard to specific embodiments. However, the benefits,

20   advantages, solutions to problems, and any element(s) that may cause any benefit, advantage, or solution to occur or become more pronounced are not to be construed as a critical, required, or essential feature or element of any or all the claims. As used herein, the terms "comprises," "comprising," or any other variation thereof, are intended to cover a non-exclusive inclusion, such that a

25   process, method, article, or apparatus that comprises a list of elements does not

include only those elements but may include other elements not expressly listed or inherent to such process, method, article, or apparatus.